SG6859A — Low Cost, Green-Mode PWM Controller for Flyback Converters

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SG6859A Low-Cost, Green-Mode PWM Controller for Flyback Converters

Features

- Green-Mode PWM
- Supports the "Blue Angel" Standard
- Low Startup Current: 9 µA
- Low Operating Current: 3 mA
- 300 mA Driving Capability
- Leading-Edge Blanking
- Constant Output Power Limit
- Universal Input
- Built-in Synchronized Slope Compensation
- Current-Mode Operation
- Cycle-by-cycle Current Limiting
- Under-Voltage Lockout (UVLO)
- Programmable PWM Frequency with Frequency Hopping
- V_{DD} Over-Voltage Protection (Auto Restart)
- Gate Output Voltage Clamped at 17 V
- Low Cost
- Few External Components Required
- Small SOT23-6L Package

Applications

General-purpose switching mode power supplies and flyback power converters, such as:

- Battery chargers for cellular phones, cordless phones, PDAs, digital cameras, and power tools
- Power adapters for ink jet printers, video game consoles, and portable audio players
- Open-frame SMPS for TV/DVD standby and auxiliary supplies, home appliances, and consumer electronics
- Replacements for linear transformers and RCC SMPS
- PC 5 V standby power

Description

This highly integrated PWM controller provides several enhancements designed to meet the low standby-power needs of low-power SMPS. To minimize standby power consumption, the proprietary green-mode function provides off-time modulation to linearly decrease the switching frequency under light-load conditions. This green-mode function enables the power supply to meet even the strictest power conservation requirements.

The BiCMOS fabrication process enables reducing the startup current to $9 \ \mu$ A and the operating current to $3 \ m$ A. To further improve power conservation, a large startup resistance can be used. Built-in synchronized slope compensation ensures the stability of peak current mode control. Proprietary internal compensation provides a constant output power limit over a universal AC input range ($90 \ V_{AC}$ to $264 \ V_{AC}$). Pulse-by-pulse current limiting ensures safe operation during short-circuits.

To protect the external power MOSFET from damage by supply over voltage, the SG6859A's output driver is clamped at 17 V. SG6859A controllers can be used to improve the performance and reduce the production cost of power supplies. The best choice for replacing linear and RCC-mode power adapters, the SG6859A is available in 8-pin DIP and 6-pin SOT23-6L packages.

| Ordering Information | | | | | |
|----------------------|-----------------------------|----------|----------------|--|--|
| Part Number | Operating Temperature Range | Package | Packing Method | | |
| SG6859ATZ | | SOT23-6L | | | |
| SG6859ATY | -40 to +105°C | SOT23-6L | Tape & Reel | | |
| SG6859ADZ | | | Tuba | | |
| SG6859ADY | | DIP-8 | Tube | | |

Application Diagram



Figure 1. Typical Application





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Pin Configurations







Figure 5. DIP-8 Pin Configuration

Pin Definitions

| DIP Pin # | SOT Pin # | Name | Description | |
|--------------|--------------|-------|---|--|
| 1 | 6 | GATE | The totem-pole output driver for driving the power MOSFET. | |
| 2 | 5 | VDD | Power supply | |
| 3 | | NC | No connection | |
| 4 | 4 | SENSE | Current sense. This pin senses the voltage across a resistor. When the voltage reaches the internal threshold, PWM output is disabled. This activates over-current protection. This pin also provides current amplitude information for current-mode control. | |
| 5 | 3 | RI | A resistor connected from the RI pin to ground generates a constant current source used to charge an internal capacitor and determine the switching frequency. Increasing the resistance reduces the amplitude of the current source and the switching frequency. A 95 k Ω resistor R _I results in a 50 μ A constant current I _I and a 70 kHz switching frequency. | |
| 6 | | NC | No connection | |
| 7 | 2 | FB | Feedback. The FB pin provides the output voltage regulation signal. It provides feedback to the internal PWM comparator, so that the PWM comparator can control the duty cycle. | |
| 8 | 1 | GND | Ground | |

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

| Symbol | Parameter | | | Max. | Unit |
|--------------------|---|---------|------|------|------|
| V_{VDD} | DC Supply Voltage ^(1, 2) | | | 30 | V |
| V _{FB} | Input Voltage to FB Pin | | -0.3 | 7.0 | V |
| V _{SENSE} | Input Voltage to Sense Pin | | -0.3 | 7.0 | V |
| TJ | Operating Junction Temperature | | | 150 | °C |
| | | OT23-6L | | 273 | °C/W |
| Θ_{JA} | A Thermal Resistance (Junction-to-Air) DIP | | | 113 | °C/W |
| T _{STG} | Storage Temperature Range | | -55 | +150 | °C |
| ΤL | Lead Temperature (Wave Soldering or IR, 10 Seconds) | | | +260 | °C |
| _ | Electrostatic Discharge Capability, Human Body Model JESD22-A114 | | | 3.5 | kV |
| ESD | Electrostatic Discharge Capability, Charged Device Model JESD22-C101 1.5 kV | | | kV | |
| | Electrostatic Discharge Capability, Machine Model, JESD22-A115 | | | 200 | V |

Notes:

1. All voltage values, except differential voltages, are given with respect to GND pin.

2. Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device.

Electrical Characteristics

Unless otherwise noted, V_{DD} =15 V and T_A =25°C.

| Symbol | Parameter | | Conditions | Min. | Тур. | Max. | Units |
|--------------------------|---|---------------------------|--|------|-----------------------------|------|-------|
| V _{DD} Sectior | 1 | | | | | 1 | |
| V _{DD-OP} | Continuously Ope | eration Voltage | | | | 22 | V |
| V _{DD-ON} | Turn-on Threshol | d Voltage | | 15.5 | 16.5 | 17.5 | V |
| $V_{\text{DD-OFF}}$ | Turn-off Threshol | d Voltage | | 8.5 | 9.5 | 10.5 | V |
| I _{DD-ST} | Startup Current | | V _{DD} =V _{DD-ON} – 0.1 V | | 9 | 15 | μA |
| I _{DD-OP} | Operating Supply | Current | V _{DD} =15 V, C _L =1 nF | | 3.0 | 3.5 | mA |
| V _{DD-OVP} | V _{DD} Over-Voltage | Protection Level | Auto Restart | 24 | 25 | 26 | V |
| t _{D-VDDOVP} | V _{DD} Over-Voltage | Protection Debounce | Auto Restart | | 125 | | μs |
| $V_{\text{DD-G OFF}}$ | V _{DD} Low-Thresho Green-off Mode | ld Voltage to Exit | | | V _{DD-} OFF + 1 | | V |
| Feedback I | nput Section | | | | | | |
| Z _{FB} | Input Impedance | | | | 5 | | kΩ |
| V _{FB-OPEN} | FB Output High V | oltage | | 5 | | | V |
| V _{FB-OL} | FB Open-loop Tri | gger Level | | 4.3 | 4.6 | 4.9 | V |
| t _{D-OLP} | Delay of FB Pin C | pen-loop Protection | | | 56 | | ms |
| V _{FB-N} | Green-Mode Entr | y FB Voltage | | 2.60 | 2.85 | 3.10 | V |
| V_{FB-G} | Green-Mode End | ing FB Voltage | | | 2.2 | | V |
| S _G | Green-Mode Mod | ulation Slope | R _I =95 kΩ | 40 | 75 | 100 | Hz/m |
| Current-Se | nse Section | | | | | | |
| ZSENSE | Input Impedance | | | 10 | | | kΩ |
| t _{PD} | Delay to Output | | | 40 | 55 | 100 | ns |
| V _{STHFL} | Flat Threshold Vo | oltage for Current Limit | 9 | | 1 | | V |
| VSTHVA | Valley Threshold | Voltage for Current Limit | 1 | 0.75 | 0.80 | 0.85 | V |
| t _{LEB} | Leading-Edge Bla | anking Time | | 270 | 320 | 370 | ns |
| DCY _{SAW} | Duty Cycle of SA | W Limit | | 1 | 40 | | % |
| Oscillator S | Section | | | | | | |
| | | Center Frequency | | 65 | 70 | 75 | kHz |
| fosc | Frequency | Hopping Range | R _I =95 kΩ | | ±4.9 | | |
| T _{HOP} | Hopping Period | | R _I =95 kΩ | | 3.7 | / | ms |
| fosc-g | Green-Mode Free | quency | R _I =95 kΩ | | 20 | | kHz |
| f _{DV} | Frequency Variation vs. V _{DD} Deviation | | V _{DD} =13.5 to 22 V | 0 | 0.02 | 2.00 | % |
| f _{DT} | Frequency Variat | ion vs. Temp. Deviation | T _A =-20 to 85°C | | | 2 | % |
| Output Sec | tion | | | / | 1 | | |
| DCY _{MAX} | Maximum Duty C | ycle | | 62 | 67 | 72 | % |
| V _{GATE-L} | Output Voltage L | - | V _{DD} =15 V, I _O =20 mA | | | 1.4 | V |
| V _{GATE-H} | Output Voltage H | | V _{DD} =13.5 V, I _O =20 mA | 8 | 1 | | V |
| tr | Rising Time | - | V _{DD} =15 V, C _L =1 nF | | 150 | | ns |
| t _f | Falling Time | | V _{DD} =15 V, C _L =1 nF | 1 | 55 | | ns |
| V _{GATE} -CLAMP | Ũ | ltage | V _{DD} =22 V | 16 | 17 | 18 | V |





Operation Description

SG6859A devices integrate many useful designs into one controller for low-power, switch-mode power supplies. The following descriptions highlight some of the features of the SG6859A series.

Startup Current

The startup current is only 9 μ A. Low startup current allows a startup resistor with high resistance and low-wattage to supply the startup power for the controller. A 1.5 M Ω , 0.25 W, startup resistor and a 10 μ F/25V V_{DD} hold-up capacitor are sufficient for an AC-to-DC power adapter with a wide input range (100 V_{AC} to 240 V_{AC}).

Operating Current

The operating current has been reduced to 3 mA. The low operating current results in higher efficiency and reduces the V_{DD} hold-up capacitance requirement.

Green-Mode Operation

The proprietary green-mode function provides off-time modulation to linearly decrease the switching frequency under light-load conditions. On-time is limited to provide stronger protection against brownouts and other abnormal conditions. The feedback current, which is sampled from the voltage feedback loop, is taken as the reference. Once the feedback current exceeds the threshold current, the switching frequency starts to decrease. This green-mode function dramatically reduces power consumption under light-load and zeroload conditions. Power supplies using the SG6859A can meet even the strictest regulations regarding standby power consumption.

Oscillator Operation

A resistor connected from the RI pin to ground generates a constant current source used to charge an internal capacitor. The charge time determines the internal clock speed and the switching frequency. Increasing the resistance reduces the amplitude of the input current and the switching frequency. A 95 k Ω R_I resistor results in a 50 µA constant current, I_I, and a 70 kHz switching frequency. The relationship between R_I and the switching frequency is:

| $f_{PWM} = \frac{6650}{R_1(k\Omega)} (kHz)$ | (1) |
|---|-----|
| 11(1122) | |

The recommended f_{PWM} is from 50kHz to 80kHz.

Leading-Edge Blanking

Each time the power MOSFET is switched on, a turn-on spike occurs at the sense-resistor. To avoid premature termination of the switching pulse, a 320 ns leadingedge blanking time is built in. Conventional RC filtering can be omitted. During this blanking period, the currentlimit comparator is disabled and cannot switch off the gate driver.

Constant Output Power Limit

When the SENSE voltage across the sense resistor, $R_{S},$ reaches the threshold voltage (around 1 V), the

output GATE drive is turned off following a short propagation delay, $t_{\text{PD}}.$ This propagation delay introduces an additional current proportional to t_{PD}V_{IN}/L_P. The propagation delay is nearly constant, regardless of the input line voltage V_{IN}. Higher input line voltages result in larger additional currents. At high input line voltages, the output power limit is higher than at low input line voltages. To compensate for this output power limit variation across a wide AC input range, the threshold voltage is adjusted by adding a positive ramp. This ramp signal rises from 0.8 V to 1 V, then flattens out at 1 V. A smaller threshold voltage forces the output GATE drive to terminate earlier, which reduces the total PWM turn-on time and makes the output power equal to that of low line input. This proprietary internal compensation ensures a constant output power limit for a wide AC input voltage range (90 V_{AC} to 264 V_{AC}).

Under-Voltage Lockout (UVLO)

The turn-on and turn-off thresholds are fixed internally at 16.5 V and 9.5 V. During startup, the hold-up capacitor must be charged to 16.5V through the startup resistor to enable the SG6859A. The hold-up capacitor continues to supply VDD until power can be delivered from the auxiliary winding of the main transformer. V_{DD} must not drop below 9.5 V during this startup process. This UVLO hysteresis window ensures that hold-up capacitor is adequate to supply V_{DD} during startup.

Gate Output

The BiCMOS output stage is a fast totem pole gate driver. Cross conduction has been avoided to minimize heat dissipation, increase efficiency, and enhance reliability. The output driver is clamped by an internal 17 V Zener diode to protect power MOSFET transistors against undesired over-voltage gate signals.

Built-in Slope Compensation

The sensed voltage across the current-sense resistor is used for current-mode control and pulse-by-pulse current limiting. Built-in slope compensation improves stability and prevents sub-harmonic oscillations due to peak-current mode control. The SG6859A has a synchronized, positively-sloped ramp built-in at each switching cycle. The slope of the ramp is:

Duty (max.)

(2)

Noise Immunity

Noise from the current sense or the control signal can cause significant pulse-width jitter, particularly in continuous-conduction mode (CCM). While slope compensation helps alleviate these problems, further precautions should still be taken. Good placement and layout practices should be followed. Avoiding long PCB traces and component leads, locating compensation and filter components near the SG6859A, and increasing power MOS gate resistance improve performance.



Figure 16. Reference Circuit

Bill of Materials (BOM)

| Reference | Component | Reference | Component |
|----------------|----------------------|----------------|------------------|
| BD1 | BD 1 A/500 V | L2 | 10 µH 6 mm |
| CX1 (Optional) | XC 0.1 µF | Q1 | MOSFET 1 A/600 V |
| CY1 (Optional) | YC 1nF/400 V (Y1) | R1,R2 | R 750 KΩ 1206 |
| C1 | CC 10 nF/500 V | R3,R4 | R 47 KΩ 1206 |
| C2 | EC 10 µF/400 V 105°C | R5 | R 47 Ω 1206 |
| C3 | CC 1 nF/500 V | R6 | R 4.7 Ω 1206 |
| C4 | EC 10 μF/50 V | R7 | R 100 Ω 0805 |
| C6 | CC 4.7 nF 0805 | R8 | R 10 Ω 1206 |
| C7 (Optional) | CC 1 nF/100 V 1206 | R9 | R 100 KΩ 0805 |
| C8 | EC 470 µF/10 V 105°C | R10 (Optional) | R 10 Ω 1206 |
| C9 | EC 220 µF/10 V 105°C | R11 | R 100 Ω 1/8W |
| C10 | CC 2.2 nF 0805 | R12 | R 33 KΩ 0805 |
| C11 | N.C. | R13 | R 33 KΩ 1/8W |
| D1 | Diode FRI07 | R14 | R 4.7 KΩ 0805 |
| D2 | Diode FR102 | R15 | R 0 Ω 0805 |
| D4 | Diode SB360 | T1 | EE-16 |
| D5 (Optional) | ZD 6.8 V 0.5 W | U1 | IC SG6859A |
| F1 | R 1 Ω/0.5 W | U2 | PC817 |
| L1 | 20 mH 6•8 mm | U3 | TL431 |





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